



UNIK4230: Mobile Communications

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Propagation characteristis of wireless channel - I





Agenda

- Radio wave propagation phenomena
 - Reflection
 - Refraction
 - Diffraction
 - Scattering
- Signal attenuation
 - Attenuation and fading
 - Path loss
 - Hata model
- Indoor propagation
- More on fading





Agenda

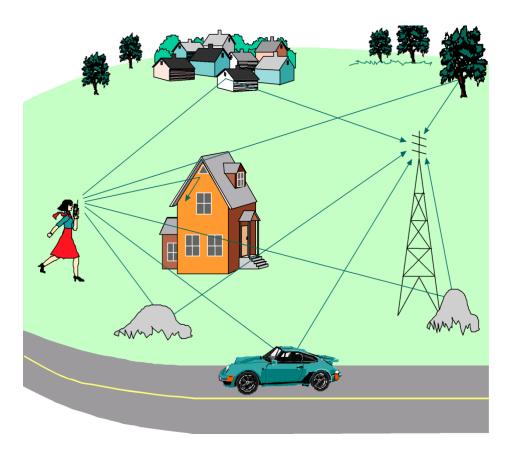
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Radio channel

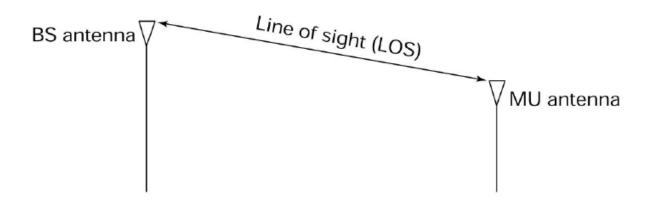
- The transmitted signal arrives at the receiver from different directions at different times over a number of ways
 - Line of sight or non-line of sign







Line of sight propagation

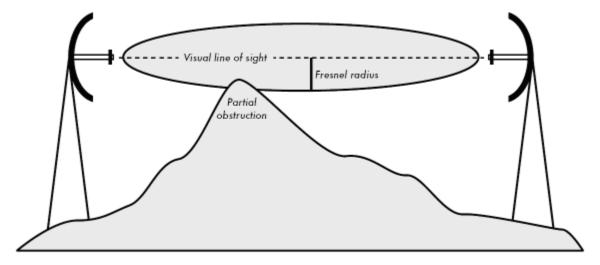


- Unobstructed line of sight propagation between the transmitter and receiver
 - Lower attenuation
 - There must be no obstruction on LOS
 - Objects not even on direct LOS can interfere radio transmission (see Fresnel Zone)





Fresnel zones



$$F_n = \sqrt{\frac{n\lambda d_1 d_2}{d_1 + d_2}}$$

All units are meter

1st Fresnel zone while D in Km and f in GHz

$$r = 8.657\sqrt{\frac{D}{f}}$$

The Fresnel zone is partially blocked on this link, although the visual line of sight appears clear..

- Fresnel zones determine whether a given obstacle will cause a constructive or desctructive interference at the receiver due to reflection
 - Reflection can enhance received signal if reflected and direct signals arrive in-phase
 - Its important to clear obstruction from first Fresnel zone

The radius of first Fresnel zone, r = 17.31 * sqrt(N(d1*d2)/(f*d)) ...where r is the radius of the zone in meters, N is the zone to calculate, d1 and d2 are distances from obstacle to the link end points in meters, d is the total link distance in meters, and f is the frequency in MHz.





Fresnel zones

 For example, let's calculate the size of the first Fresnel zone if the first Fresnel zone in the middle of a 2km link, transmitting at 2.437GHz (802.11b channel 6):

```
r = 17.31 sqrt(1 * (1000 * 1000) / (2437 * 2000))
r = 17.31 sqrt(1000000 / 4874000)
r = 7.84 meters
```

Assuming both of our towers were ten meters tall, the first Fresnel zone would pass just 2.16 meters above ground level in the middle of the link.



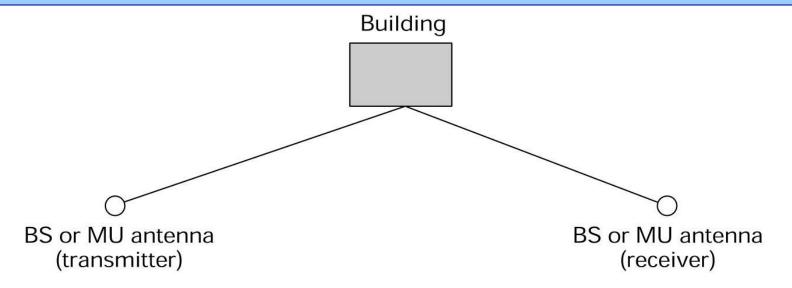


Fresnel zone: exercise

 How tall could a structure in the middle of a 2km point be to clear 60% (N=0.6) of the first Fresnel zone? Transmitter is transmitting at 2.437GHz (802.11b channel 6)



Reflection

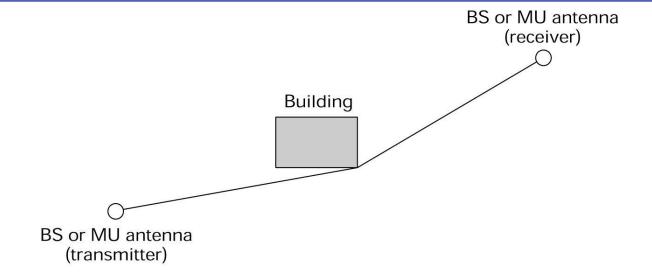


- Reflection occurs when a radio wave strikes a barrier with a dimension larger than the wave length of the wave
 - E.g. Buildings, ground, vehicles etc.





Diffraction

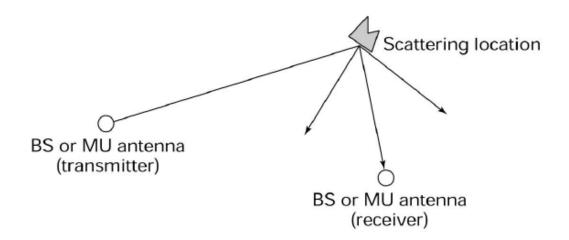


- Diffraction occurs when the radio wave hits an obstacle with sharp irregularities, edge, small gap
 - Size of the object (e.g. edge) must be comparable to or smaller than the wavelength of the radio wave
 - E.g. Bending around the object (typically corner of the houses or hills)





Scattering



- Scattering occurs when the radio wave strikes the obstacles with dimension smaller than the wave length of the radio wave
 - E.g. Vegetation, street signs etc.





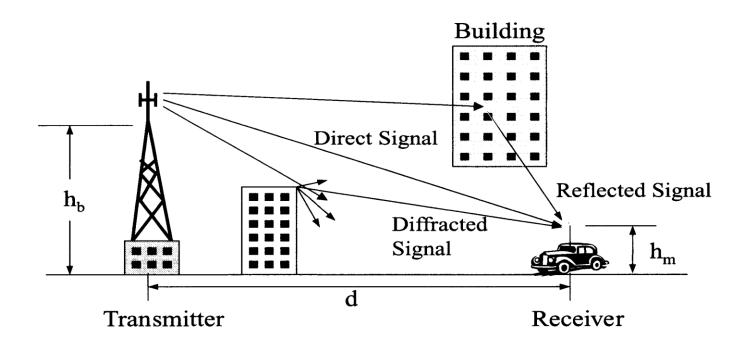
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Attenuation and fading







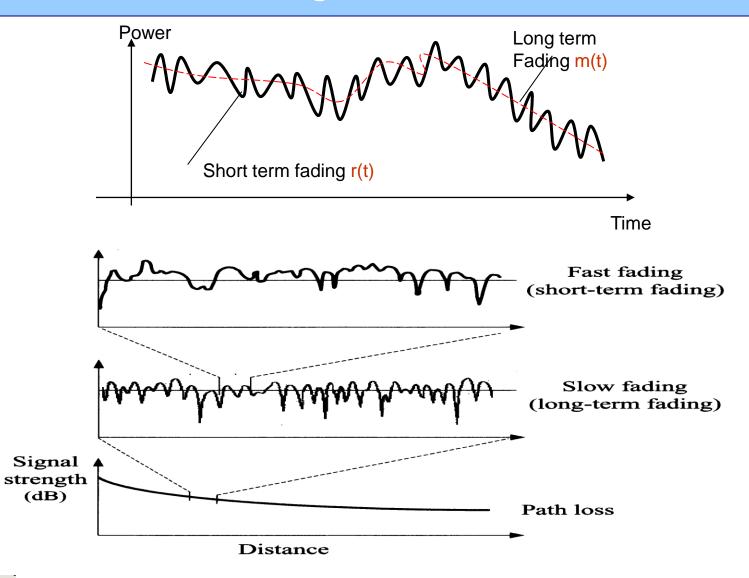
Attenuation and fading

- Fading is devation of attenuation a radio wave experience over certain propagation media.
 - Distance dependent attenuation
 - Fast fading: Rapid fluctuation of signal over a small areas. Fast fading occurs due to multipath propagation
 - Fast fading is caracterized by Rayleigh and Rician distribution.
 - Rayleigh distribution: It assumes infinite reflected path with all possible attenuation and no direct path. E.g. It is used characterize worst case urgan or indoor communications
 - Rician distribution: It assumes a direct path from TX to RX as well as infinite reflected paths. E.g. Used to characterize satellite communication channels
 - Slow fading: It is long-term fading effect caused by large obstruction (shadowing) such as large building or hills
 - Shadowing is modeled using using log-normal distribution.





Attenuation and fading







About the term dB

- Widely used to measure e.g. gain, attenuation, signal to noise ratio (SNR) etc.
- The ratio of power value P_a to another power value P_b is calculated as:

$$X_{dB} = 10\log_{10}\left(\frac{P_a}{P_b}\right) dB$$

Example:

Ratio =
$$0.1 = -10 \text{ dB}$$

= $1 = 0 \text{ dB}$
= $10 = 10 \text{ dB}$
= $100 = 20 \text{ dB}$

Decibel (dB) is a dimensionless Unit

$$P_a = 10^{\frac{X_{dB}}{10}} P_b$$
 watt given P_b in watt





About the term dBm

- dBm (decibel-milliwatt) is the power unit in dB referenced to 1 mW.
- It measures absolute power in radio, microwave and fiber option network.
 - dBm can measure both very small and very large values in short form
- To measure an arbitrary power P_a as x dBm:

$$x = 10\log_{10}\left(\frac{P_a(mW)}{1mW}\right) \quad dBm$$

Example:

 $P_a = 1 \text{ mW}, x = 0 \text{ dBm}$

 $P_a = 1 \text{ W}$, x = 30 dBm, maximum outout power of GSM 1800 mobile phone

 $x = 33 \text{ dBm}, P_a = 2 \text{ W}$

x = 80 dBm, $P_a = 100$ KW, P_{tx} of FM radio transmitter with 50km range

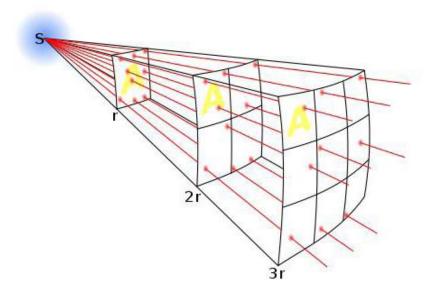
dBm is an absolute measure of power in mW





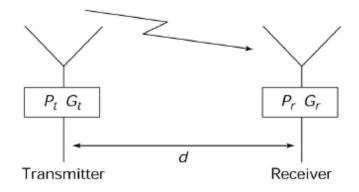
 When there is line of sight between transmitter and receiver, received power follows inverse square law:

$$P_r \propto d^{-2}$$





Received power in free space can be expressed as:



$$P_r = \frac{P_t G_t G_r \lambda^2}{4\Pi^2 d^2 L}$$

 $P_{t} = transmit\ power$

 G_t , G_r = gain of transmitter, receiver antenna

 $L = other\ losses\ (e.g.\ filtertap\ , antennatap\)$





Free space path loss:

$$L = \left(\frac{4\Pi d}{\lambda}\right)^{2} = \left(\frac{4\Pi df}{c}\right)^{2}$$

$$d = dis \tan ce \text{ from transmitter } (m)$$

$$\lambda = signal \text{ wavelength } (m)$$

$$f = signal \text{ frequency } (Hz)$$

$$c = speef \text{ of light}$$

$$L_{dB} = 32.44 + 20\log_{10}(f) + 20Log_{10}(d)$$

where d in km and f in MHz





 If received power is known in a reference distance d_{ref}, received power in an arbritary distance can be calculated:

$$P_r(d) = P_r(d_{ref}) \left(\frac{d_{ref}}{d}\right)^2$$

In dBm

$$P_r(d)[dBm] = 10\log_{10}(P_r(d_{ref})) + 20\log_{10}\left(\frac{d_{ref}}{d}\right)$$





Path loss

If transmitted and received power are known, path loss can be calculated:

$$L(dB) = 10\log_{10}\left(\frac{P_t}{P_r}\right)$$

if both transmitter and receiver has no gain, its identical to free space loss

From the above equation, we can also write:

Path Loss (dB) = Transmit Power (dBm) - Received Power (dBm)





Attenuation factor

- In real case attenuation is much higher because signal propagation path is not really free space.
- With attenuation factor, received power:

$$P_r \propto d^{-\nu}$$

v=2 for free space

Typical values for urban areas are 3-5

With reference distance one can write:

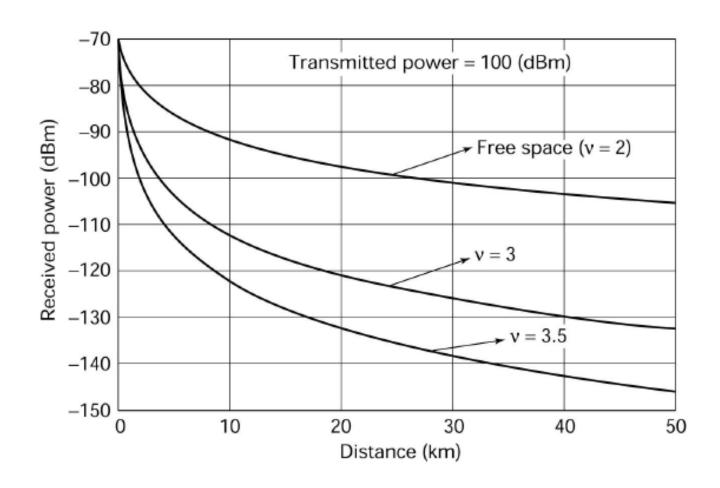
$$P_r(d)[dBm] = 10\log_{10}(P_r(d_{ref})) + 10.v.\log_{10}\left(\frac{d_{ref}}{d}\right)$$





Attenuation factor

Received as a function of distance for different values of v:







About channel model

- Models are mathemetical description of attenuation that are used for system design, system simulation or radio planning purposes
- Two types of channel model:
 - Empirical model: developed based on large collection of data for a specific scenario (e.g. Urban, sub-urban); do not point out exact behavior rather most likely behavior of the channel
 - Analytical model: takes into account link specific geometri (e.g. curve of hills, edges, big buildings etc.)
- In Practice often takes combination with site-specific correction factors used in addition to empirical models





Okumura model

- Combining all these causes (reflection, scattering, and diffraction), Okumura et al. (1968) proposed channel model
- The model includes correction factor to account for terrain
- But correction factors have to be incorporated for every scenario
 - Hata (1980) proposed a model to overcome the problem

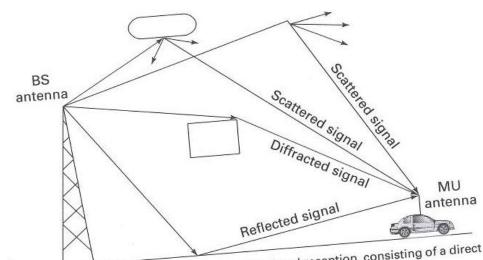


FIGURE 2.10 The most general case of signal reception, consisting of a direct path, a reflected path, a scattered path, and a diffracted path.

. . MII through t





• In Hata model, path loss in urban areas is given by:

$$\begin{split} L_p(dB) &= 69.55 + 26.16 \log_{10}(f_0) + (44.9 - 6.55 \log_{10}h_b) \log_{10}d - 13.82 \log_{10}h_b - a(h_{mu}) \\ where \\ f_0 &= carrier \ frequency \ (MHz) \\ d &= separation \ between \ BTS \ and \ MU \ (km); \ d \geq 1 \ km \\ h_b &= height \ of \ the \ BTS \ antenna \ (m) \\ h_{mu} &= height \ of \ the \ MU \ antenna \ (m) \\ a(h_{mu}) &= correction \ factor \ for \ MU \ antenna \ height \end{split}$$

• For large cities, the correction factor a(h_{mu}) is given by:

$$a(h_{mu}) = 3.2[\log_{10}(11.75h_{mu})]^2 - 4.97$$
 $f_0 \ge 400 MHz$

• For small and medium cities, the correction factor a(h_{mu}) is given by:

$$a(h_{mu}) = [1.1\log_{10}(f_0) - 0.7]h_{mu} - [1.56\log_{10}(f_0) - 0.8]$$





For suburban and rural areas following correction factors are used:

$$L_{sub}(dB) = L_p - 2[\log_{10}(\frac{f_0}{28})]^2 - 5.4$$

$$L_{rur}(dB) = L_p - 4.78[\log_{10}(f_0)]^2 + 18.33\log_{10}f_0 - 40.94$$
 where L_p is the loss in small – to – medium citites .

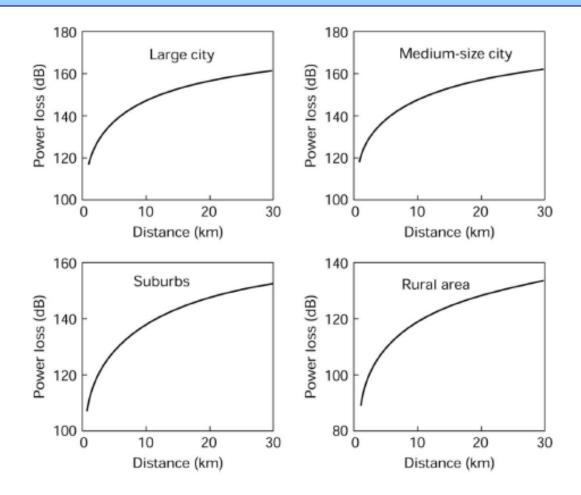


Figure shows loss calculation based on hata model for four different environments given that f_0 =900MHz, h_b =150m, h_{mu} =1.5m





 COST 231 extension or extension of hata model to PCS (personal communication system):

 $L(dB) = 46.3 + 33.93 \log_{10}(f_0) - 13.82 \log_{10}(h_b) - a(h_{mu}) + [44.9 - 6.55 \log_{10}(h_b)] \log_{10}d - Corr.$ where

Corr. is the additional correction factor given by

Corr. = 0 dB for medium city and suburban areas

Corr. = 3 dB for metropoli tan areas

This model is valid for the following parameter:

 $f_0:1500-2000MHz$

 $h_b: 30-200m$

 $h_{mu}:1-10m$

d:1-20km





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Indoor propagation model

- Models so far presented are not sufficient to predict signals in indoor
- Indoor propagation sees reflect, scatter, and diffract due to walls, ceilings, furnitures etc. (i.e. many obstacles)
- Best approach to model indoor: classify these environments into different `zone' configurations
 - Extra large zone
 - Large zone
 - Middle zone
 - Small & microzone





Indoor propagation model

Extra large zone:

- A BTS outside building takes all the traffic in the buildings
- Loss = path-dependent losses (from BTS to building) + penetrationdependent losses (penetration of various floors & walls)

Large zone:

- Large buildings with small density of users
- The building is covered by a single indoor BTS located within the building itself
- General formula of path loss can be used. L(d) = L0 (d0) [d0/d]v
- Loss is determined whether users are in the same floor as BTS
 (attenuation factor 2-3 if Tx and Rx on the same floor, it will be greater
 than 3 if they are on different floors)





Indoor propagation model

Middle zone:

- Bulding structure is large and heavily populated (e.g. shopping malls)
- A number of BTSs serve the users
- Loss = Free space path loss + floor loss + wall loss + reflection loss

Small zone and microzone:

- Buildings having many walls and partitions
- Loss depends on the material of the walls and partitions
- Need the provision of one BTS for each room
- Usually heavy traffic in each room
- Large-zone model can be used with appropriate path loss exponent: v=2 for LOS, v>2 for NLOS.





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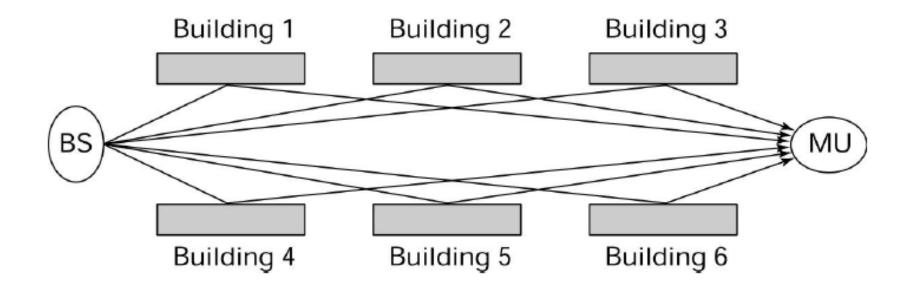
Fading

- In addition to propagation loss, attenuation may also fluctuate with position and time
- Propagation fluctuates around mean value
- Fading describes this signal fluctuation around mean value
- Primary cause of fading is signal traversing multiple path
- Fading can be described in three ways:
 - Multipath
 - The statistical distribution of the received signal envelope (e.g. Rayleigh)
 - Duration of fading (e.g. long-term, short-term)





 Signal leaves the transmitting antenna and can take different paths to reach the receiver (due to reflection, deffraction, scattering etc.)







- Signal components arrive at receive antenna are independent of each other
 - Hence, signal received at the antenna can be expressed as the vector sum of the signal components
- Assuming Rx stationary & no direct path exists (Tx-Rx), the received signal e_r(t):

$$e_r(t) = \sum_{i=1}^{N} a_i p(t - t_i)$$

where

 $a_i = amplitude of the received component i$

p(t) = transmitted signal shape

 t_i = time taken by signal component i to reach receiver

 $N = Number\ of\ paths\ taken\ by\ the\ signal$





 Instead of using sum of delayed components, received signal can also showed using phasor notation:

$$e_r(t) = \sum_{i=1}^{N} a_i \cos(2\Pi f_0 t + \phi_i)$$

where

 $a_i = amplitude of the received component i$

 $f_0 = carrier frequency$

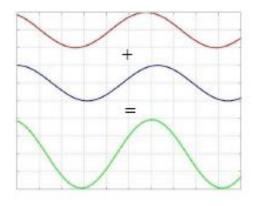
 $\phi_i = Phase \ of \ i^{th} \ signal \ component$

N = Number of paths taken by the signal

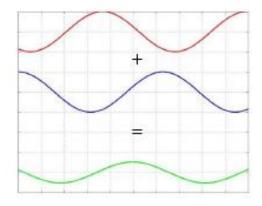




- Resulting signal is the random summation of different signals (cosine shaped signals)
 - Leads to a random variation depending on the relative phase between signal components
 - Creates constructive and destructive summation.



Constructive sum



Destructive sum





Rayleigh

$$e_r(t) = \sum_{i=1}^{N} a_i \cos(2\Pi f_0 t + \phi_i)$$

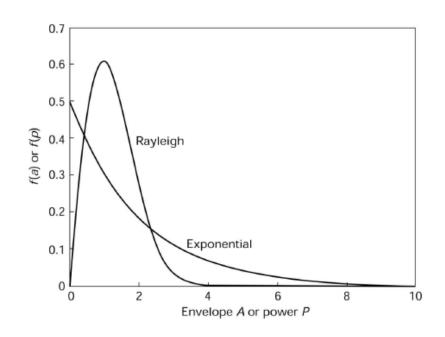
$$= \cos(2\Pi f_0 t) \sum_{i=1}^{N} a_i \cos(\phi_i) - \sin(2\Pi f_0 t) \sum_{i=1}^{N} a_i \sin(\phi_i)$$

$$= X\cos(2\Pi f_0 t) - Y\sin(2\Pi f_0 t)$$

where

$$X = \sum_{i=1}^{N} a_i \cos(\phi_i), \quad Y = \sum_{i=1}^{N} a_i \sin(\phi_i)$$

- X and Y are independent and identically distributed Gaussian random variable
- Under this condition envelop of the received signal A, given by (X²+Y²)^{1\2}, will be Rayleigh distributed







Rayleigh and Rician Model

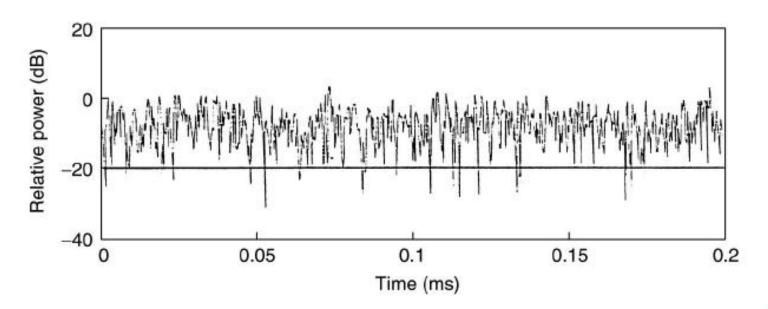
- Rayleigh fading model assumes there is no line of sight (LOS) or most applicable when there is no dominant propagation along the LOS.
- Rician fading occurs when one of the paths, typically a line of sight signal is much stronger than the others. That means it assumes a LOS
- Hence, Rayleigh model can be also considered a special case of Rician model





Outage

- Every receiver is designed to operate at an acceptable level only if a certain minimum power, P_{thr}, is being received
- The receiver will be in outage whenever power goes below this threshold value
- Outage is the implication of fading; following system goes into outage if the threshold is set to -20 dB of relative power

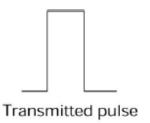


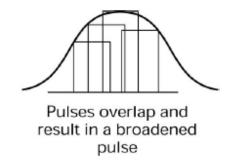




Multipath and Intersymbol interference

- Fading may affect the shape of the received signal pulse
- Figure: four different paths, at the receiver pulse arrives at four different times
- Envelope of the overlapping pulse showed a broadened pulse – leads to intersymbol interference (ISI)

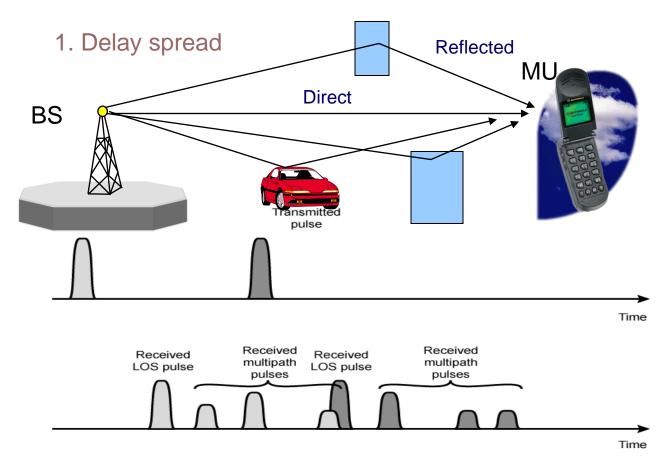








Multipath and Intersymbol interference



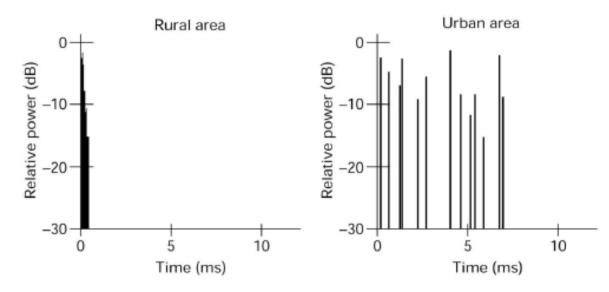
- Intersymbol interference (ISI) occurs if the delay spread of the channel exceeds the symbol time (or the sampling interval)
- Cancellation of ISI is done via an equalizer at the receiver





Impulse response

- Impulse corresponding to multiple paths arrive at the receiver at different times and with different power depending on the nature of the channel (e.g. reflection, defraction, scattering etc.)
- These arrival times of signal with different powers can be used to define the impulse response of the channel
- Figure:
 - rural areas due to fewer tall structure, multiple paths are closed to each other
 - Urban areas multiple paths are more diversed and received signals are spread out







Symbol rate and bandwidth

 There is a direct correlation between symbol rate, R (symbol/s) and information bandwidth, B_s (Hz) in a radio connection:

$$R \propto B_s$$

Means,

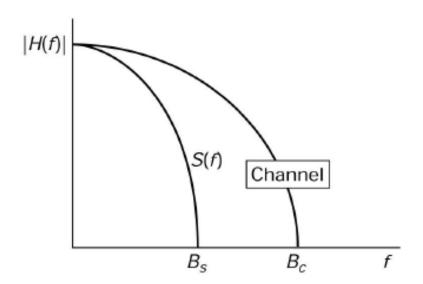
- High symbol rate (bit rate) -> high bandwidth (broadband)
- Low symbol rate (bit rate) -> low bandwidth (narrowband)





Flat fading channel

- If the channel bandwidth B_c is larger than message bandwidth B_s, all the frequecy components in the message will arrive at the receiver with little or no distortion
- ISI will be neligible
- The channel will be defined as flat fading channel
- Rural areas can be characterized as nearly flat fading channel

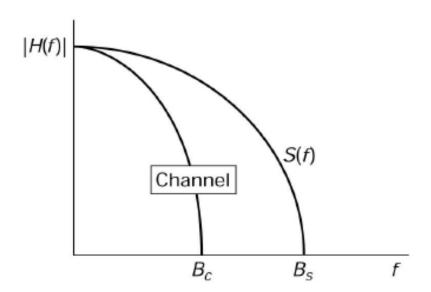






Frequency selective channel

- If the message bandwidth B_s is larger than channel bandwidth B_c, different frequecy components in the message will arrive at the receiver at different time
- Resulting pulse broadening ISI
- The channel is classified as frequency selective channel
- The flat fading channel can become frequency selective channel if the information is transmitted with higher and higher bandwidth







Doppler effect

- So far, we assumed mobile phone being stationary
- The motion of the mobile unit results a doppler shift in the frequency of the received signal
- The maximum doppler shift is expressed as,

$$f_d = f_0 \frac{v}{c}$$
where
$$c = velocity \ of \ electromagnetic \ wave \ in \ free \ space$$

$$v = velocity \ of \ the \ mobile \ unit(m/s)$$

$$f_0 = frequency \ of \ the \ signal$$





Doppler effect

 Taking all the direction into account, the instantaneous frequency of the doppler shifted signal is:

$$f_{in} = f_0 + f_d \cos(\theta)$$

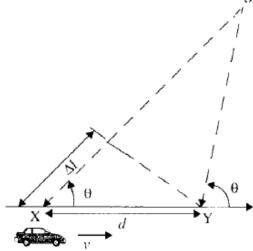


Illustration of Doppler effect.

Doppler effect

This leads to two effects:

- A level of variation at a rate that depends on the speed of the mobile
- A frequency distortion resulting from the doppler shift of the individual multi-path components (arriving at different angle)

